# **EM Scattering**

## Homework assignment 2

#### Problem 1:

Consider a dielectric layer in free space. The layer has a relative dielectric constant of  $\varepsilon_d=2.25$ , is infinite, and is parallel to the x-y plane. A plane wave coming from below the layer hits it at an angle equal to  $\theta_i$ . The polarization of the incident wave is either TE<sup>z</sup> where the electric field has no normal component (e.g., in the y-direction) or TM<sup>z</sup> where the magnetic field has no normal component. The frequency of the wave is  $f=100 {\rm GHz}$ .

- 1. Plot the magnitude of the reflection coefficient as function of  $\theta_i$  in both TE and TM cases when the layer is 1 mm-thick
- 2. Plot the magnitude of the reflection coefficient as function of  $\theta_i$  in both TE and TM cases when the layer is 5 mm-thick

#### Solution

For the TE case:

$$R_{TE} = \frac{R_{TE}^{10} \left[ 1 - \exp\left(-2jk_{1,z}d\right) \right]}{1 - \left(R_{TE}^{10}\right)^{2} \exp\left(-2jk_{1,z}d\right)} = \frac{\left(k_{0,z}^{2} - k_{1,z}^{2}\right)}{\left(k_{0,z}^{2} + k_{1,z}^{2}\right) - 2jk_{0,z}k_{1,z}\cot\left(k_{1,z}d\right)}$$

$$k_{0,z} = k_{0}\cos\theta_{i} , k_{1,z} = k_{0}\sqrt{\epsilon_{d} - \sin^{2}\theta_{i}} , \nu(\theta_{i}) \equiv \sqrt{\epsilon_{d} - \sin^{2}\theta_{i}}$$

$$\Rightarrow R_{TE} = \frac{\left(1 - \epsilon_{d}\right)}{\epsilon_{d} + \cos\left(2\theta_{i}\right) - 2j\nu(\theta_{i})\cos\theta_{i}\cot\left[\nu(\theta_{i})k_{0}d\right]}$$

$$\Rightarrow \left|R_{TE}\right|^{2} = \frac{\left(1 - \epsilon_{d}\right)^{2}}{\left[\epsilon_{d} + \cos\left(2\theta_{i}\right)\right]^{2} + 4\left(\epsilon_{d} - \sin^{2}\theta_{i}\right)\cos^{2}\theta_{i}\cot^{2}\left[\nu(\theta_{i})k_{0}d\right]}$$
(1.1)

For the TM case:

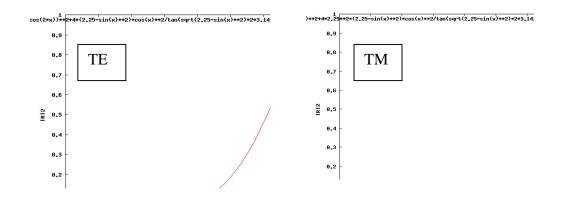
$$R_{TM} = \frac{R_{TM}^{10} \left[ 1 - \exp\left(-2jk_{1,z}d\right) \right]}{1 - \left(R_{TM}^{10}\right)^{2} \exp\left(-2jk_{1,z}d\right)} = \frac{\epsilon_{1}^{2}k_{0,z}^{2} - \epsilon_{0}^{2}k_{1,z}^{2}}{\epsilon_{1}^{2}k_{0,z}^{2} + \epsilon_{0}^{2}k_{1,z}^{2} - 2j\epsilon_{0}\epsilon_{1}k_{0,z}k_{1,z}\cot\left(k_{1,z}d\right)}$$

$$= \frac{\epsilon_{d}^{2}\cos^{2}\theta_{i} + \sin^{2}\theta_{i} - \epsilon_{d}}{\epsilon_{d}^{2}\cos^{2}\theta_{i} + \epsilon_{d} - \sin^{2}\theta_{i} - 2j\epsilon_{d}\nu(\theta_{i})\cos\theta_{i}\cot\left[\nu(\theta_{i})k_{0}d\right]}$$

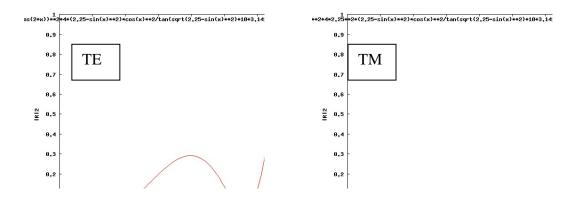
$$= \frac{\left(\epsilon_{d} - 1\right)\left[\epsilon_{d}\cos^{2}\theta_{i} - \sin^{2}\theta_{i}\right]}{\epsilon_{d}^{2}\cos^{2}\theta_{i} - \sin^{2}\theta_{i} + \epsilon_{d} - 2j\epsilon_{d}\nu(\theta_{i})\cos\theta_{i}\cot\left[\nu(\theta_{i})k_{0}d\right]}$$

$$\Rightarrow |R_{TM}|^{2} = \frac{\left(\epsilon_{d} - 1\right)^{2}\left[\epsilon_{d}\cos^{2}\theta_{i} - \sin^{2}\theta_{i}\right]^{2}}{\left[\epsilon_{d}^{2}\cos^{2}\theta_{i} - \sin^{2}\theta_{i} + \epsilon_{d}\right]^{2} + 4\epsilon_{d}^{2}\left(\epsilon_{d} - \sin^{2}\theta_{i}\right)\cos^{2}\theta_{i}\cot^{2}\left[\nu(\theta_{i})k_{0}d\right]}$$
(1.2)

For a 1 mm-thick slab:  $k_0 d = 2\pi d / \lambda = 2\pi/3$ 



For a 5 mm-thick slab:  $k_0 d = 2\pi d / \lambda = 10\pi/3$ 



#### Problem 2:

A plane radio wave is incident on the ionosphere from the air. The relative dielectric constant of the ionosphere is  $\varepsilon_d = 1 - \left( f_p / f \right)^2$  where f is the frequency of the wave and  $f_p = 5 \mathrm{MHz}$  is the plasma frequency of the ionosphere. The ionosphere is assumed to be infinitely thick.

- Compute the reflection coefficient for both TE and TM polarizations when the frequency
  of the wave is f = 3 MHz.
- 2. Repeat this calculation at the frequency of f = 6 MHz. Find the Brewster angle and the critical angle for total reflection.

#### **Solution**

TE and TM reflection coefficients:

$$R_{TE} = \frac{\cos \theta_i - \sqrt{\epsilon_d - \sin^2 \theta_i}}{\cos \theta_i + \sqrt{\epsilon_d - \sin^2 \theta_i}}, R_{TM} = \frac{\epsilon_d \cos \theta_i - \sqrt{\epsilon_d - \sin^2 \theta_i}}{\epsilon_d \cos \theta_i + \sqrt{\epsilon_d - \sin^2 \theta_i}}$$
(1.3)

1. At 3 MHz we have  $\varepsilon_d = 1 - \left(f_p / f\right)^2 = -1.77$  so that

$$R_{TE} = \frac{\cos\theta_{i} - \sqrt{\epsilon_{d} - \sin^{2}\theta_{i}}}{\cos\theta_{i} + \sqrt{\epsilon_{d} - \sin^{2}\theta_{i}}} = \frac{\cos\theta_{i} + j\sqrt{|\epsilon_{d}| + \sin^{2}\theta_{i}}}{\cos\theta_{i} - j\sqrt{|\epsilon_{d}| + \sin^{2}\theta_{i}}} \rightarrow |R_{TE}| = 1$$

$$R_{TM} = \frac{\epsilon_{d}\cos\theta_{i} - \sqrt{\epsilon_{d} - \sin^{2}\theta_{i}}}{\epsilon_{d}\cos\theta_{i} + \sqrt{\epsilon_{d} - \sin^{2}\theta_{i}}} = \frac{\epsilon_{d}\cos\theta_{i} + j\sqrt{|\epsilon_{d}| + \sin^{2}\theta_{i}}}{\epsilon_{d}\cos\theta_{i} - j\sqrt{|\epsilon_{d}| + \sin^{2}\theta_{i}}} \rightarrow |R_{TM}| = 1$$

$$(1.4)$$

2. At 6 MHz we have  $\varepsilon_d=1-\left(f_p/f\right)^2=0.306$ . Now at an angle of incidence larger than  $\arcsin\left(\sqrt{\varepsilon_d}\right)=0.586 \ \mathrm{rad}=33.6 \ \mathrm{deg}$  we have complete reflection so that  $\left|R_{TE}\right|=\left|R_{TM}\right|=1$ . Below this angle

$$R_{TE} = \frac{\cos\theta_i - \sqrt{0.306 - \sin^2\theta_i}}{\cos\theta_i + \sqrt{0.306 - \sin^2\theta_i}}, R_{TM} = \frac{0.306\cos\theta_i - \sqrt{0.306 - \sin^2\theta_i}}{0.306\cos\theta_i + \sqrt{0.306 - \sin^2\theta_i}}$$
(1.5)

The Brewster angle is

$$\arcsin\left(\sqrt{\frac{\varepsilon_d}{\varepsilon_d + 1}}\right) = 28.95 \text{ deg}$$
 (1.6)

#### Problem 3:

Consider the scalar wave equation  $(\nabla^2 + k^2)\psi = 0$  in cylindrical coordinates, and a particular solution given by

$$\psi(\rho,\phi) = \exp(-jm\phi)J_m(k\rho)$$

where k is the wave number of the medium.

- 1. Find the corresponding  $oldsymbol{M}$  and  $oldsymbol{N}$  vectors
- 2. Sketch the real and imaginary parts of their components along the radial direction (as function of  $\rho$  ) for m=1 and  $\phi$  = 0
- 3. If these vectors are to represent an electric field how does the polarization change as function of radial distance  $\rho$ ? (Again take m=1 and  $\phi$  = 0)

#### **Solution**

1. We have  $k_z = 0$ . The M and N vectors are

$$\boldsymbol{M}_{m,k_z=0} = \exp(-jm\phi) \left[ -\hat{\boldsymbol{\rho}} \frac{jm}{k\rho} J_m(k\rho) - \hat{\boldsymbol{\phi}} J'_m(k\rho) \right]$$
 (1.7)

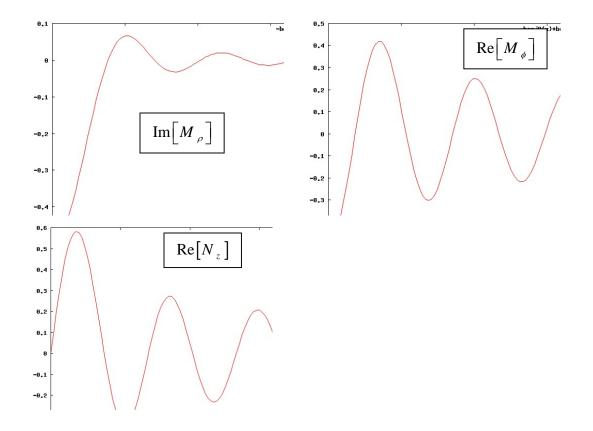
$$N_{m,k_{\perp}=0} = \exp(-jm\phi)J_m(k\rho)\hat{z}$$
(1.8)

2. For m=1 and  $\phi = 0$ 

$$\mathbf{M}_{m,k_z=0}(\rho,\phi=0) = -\hat{\rho} \frac{J}{k\rho} J_1(k\rho) - \hat{\phi} J_1'(k\rho)$$
 (1.9)

$$N_{m,k_z=0}(\rho,\phi=0) = J_1(k\rho)\hat{z}$$
 (1.10)

The radial (imaginary) and azimuthal (real) components of M and the z-component (real) of N are plotted below:



### Problem 4:

We would like to solve the 2D, **scalar** scattering problem of a scalar wave incident upon an infinitely long circular cylinder whose radius is given by a and whose axis coincides with the z-axis. The incident wave is

$$\Psi_{i}(x, y) = A \exp(-jk_{i,x}x - jk_{i,y}y)$$

with  $k_{i,x}^2 + k_{i,y}^2 = k^2$  where k is the wave number in the background medium. Like the incident wave, the scattered wave  $\Psi_s(x,y)$  satisfies the 2D wave equation

$$\frac{\partial^2 \Psi_s}{\partial x^2} + \frac{\partial^2 \Psi_s}{\partial y^2} + k^2 \Psi_s = 0$$

- Using cylindrical coordinates, find the general solution of this equation in terms of a functional series. Take into account the behavior of the scattered wave at infinity.
- 2. The boundary condition on the surface of the cylinder is  $\Psi=0$  in which  $\Psi=\Psi_i+\Psi_s$  is the total field. Using this boundary condition, solve the scattering problem by determining the coefficients of the series found above.

#### **Solution**

The scattered field may be written as

$$\Psi_{s}\left(\rho,\phi\right) = \sum_{m=-\infty}^{\infty} B_{m} \exp\left(-jm\phi\right) H_{m}^{(2)}\left(k\rho\right) \tag{1.11}$$

which satisfies the Helmholtz equation as well as the boundary condition at infinity. The plane wave may be expanded as

$$A \exp\left(-jk_{i,x}x - jk_{i,y}y\right) = A \sum_{m=-\infty}^{\infty} (-j)^m J_m\left(k\rho\right) \exp\left[-jm\left(\phi - \phi_i\right)\right]$$
(1.12)

On the surface of the cylinder  $\rho = a$  it is required that

$$\sum_{m=-\infty}^{\infty} \exp\left(-jm\phi\right) \left[ A(-j)^m J_m\left(ka\right) \exp\left(jm\phi_i\right) + B_m H_m^{(2)}\left(ka\right) \right] = 0$$
 (1.13)

As a result

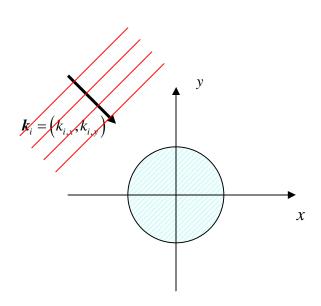
$$B_{m} = -(-j)^{m} \frac{J_{m}(ka)}{H_{m}^{(2)}(ka)} \exp(jm\phi_{i}) A$$
 (1.14)

The scattered field reads:

$$\Psi_{s}\left(\rho,\phi\right) = -A\sum_{m=-\infty}^{\infty} (-j)^{m} \frac{J_{m}\left(ka\right)}{H_{m}^{(2)}\left(ka\right)} H_{m}^{(2)}\left(k\rho\right) \exp\left[-jm\left(\phi-\phi_{i}\right)\right]$$
(1.15)

The scattered field in the far-field zone is:

$$\Psi_{s}\left(\rho,\phi\right) \sim -A\sqrt{\frac{2}{\pi k \rho}} \exp\left[-\left(jk \rho - j \pi/4\right)\right] \sum_{m=-\infty}^{\infty} \frac{J_{m}\left(ka\right)}{H_{m}^{(2)}\left(ka\right)} \exp\left[-jm\left(\phi - \phi_{i}\right)\right] (1.16)$$



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