# Probability, Random Variables, and Stochastic Processes

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# Probability

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- A random experiment is any experiment whose outcome cannot be predicted with certainty.
- A random experiment has certain outcomes  $\omega \in \Omega$ .
- The set of all possible outcomes is called the sample space Ω.
- A sample space is discrete if the number of its elements are finite or countably infinite, otherwise it is a nondiscrete sample space.
- Events are subsets of the sample space, i.e.,  $E \subset \Omega$ .
- Events are disjoint if their intersection is empty. i.e.  $E_i \cap E_j = \emptyset$ .

#### Definition (Probability Axioms)

A probability P is defined as a set function assigning nonnegative values to all events E such that

- $0 \le P(E) \le 1$  for all events.
- $P(\Omega) = 1.$

• For disjoint events  $E_1, E_2, \cdots, P(\bigcup_{i=1}^{\infty} E_i) = \sum_{i=1}^{\infty} P(E_i)$ .

- $P(E^c) = 1 P(E), \quad E^c = \Omega \setminus E.$
- $P(\emptyset) = 0.$
- $P(E_1 \cup E_2) = P(E_1) + P(E_2) P(E_1 \cap E_2).$
- $E_1 \subseteq E_2 \Rightarrow P(E_1) \leq P(E_2).$

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#### Definition (Conditional Probability)

The conditional probability of the event  $E_1$  given the event  $E_2$  is defined by

$$P(E_1|E_2) = \begin{cases} \frac{P(E_1 \cap E_2)}{P(E_2)} & , & P(E_2) \neq 0\\ 0 & , & P(E_2) = 0 \end{cases}$$

- The events  $E_1$  and  $E_2$  are said to be independent if  $P(E_1|E_2) = P(E_1)$ .
- **2** For independent events,  $P(E_1 \cap E_2) = P(E_1)P(E_2)$ .
- If the events {E<sub>i</sub>}<sup>n</sup><sub>i=1</sub> are disjoint and their union is the entire sample space, then they make a partition of the sample space Ω.
- The total probability theorem states that for an event A,  $P(A) = \sum_{i=1}^{n} P(E_i)P(A|E_i)$ .
- So Bayes's rule gives the conditional probabilities  $P(E_i|A)$  by

$$P(E_i|A) = \frac{P(E_i)P(A|E_i)}{P(A)} = \frac{P(E_i)P(A|E_i)}{\sum_{i=1}^{n} P(E_i)P(A|E_i)}$$

# **Random Variables**

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# Definition (Random Variable)

A random variable is a mapping from the sample space  $\Omega$  to the set of real numbers.

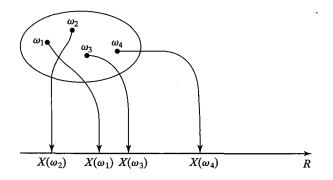


Figure: A random variable as a mapping from  $\Omega$  to  $\mathbb{R}$ .

### Definition (Cumulative Distribution Function (CDF))

The cumulative distribution function or CDF of a random variable X is defined as

$$F_X(x) = P\{\omega \in \Omega : X(\omega) \le x\} = p\{X \le x\}$$

$$0 \le F_X(x) \le 1.$$
 $F_X(-\infty) = 0, \quad F_X(\infty) = 1.$ 
 $P(a < X \le b) = F_X(b) - F_X(a).$ 

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# **Random Variables**

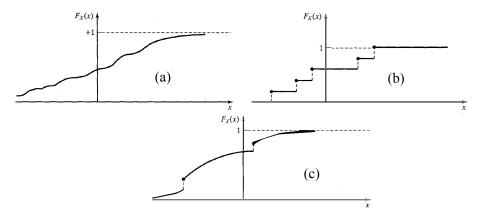


Figure: CDF for a (a) continuous (b) discrete (c) mixed random variable.

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# Definition (Probability Density Function (PDF))

The probability density function or PDF of a random variable X is defined as  $dE_{Y}(x)$ 

$$f_X(x) = \frac{dF_X(x)}{dx}$$

$$f_X(x) \geq 0.$$

 $P(a < X \le b) = \int_a^b f_X(x) dx.$ 

•  $F_X(x) = \int_{-\infty}^{x^+} f_X(u) du.$ 

# Definition (Probability Mass Function (PMF))

The probability mass function or PMF of a discrete random variable X is defined as

$$p_i = P\{X = x_i\}$$

*p<sub>i</sub>* ≥ 0.
 ∑<sub>i</sub> *p<sub>i</sub>* = 1.

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#### Statement (Bernoulli Random Variable)

The Bernoulli random variable is a discrete random variable taking two values 1 and 0, with probabilities p and 1 - p.

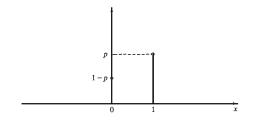


Figure: The PMF for the Bernoulli random variable.

#### Statement (Binomial Random Variable)

The binomial random variable is a discrete random variable giving the number of 1 's in n independent Bernoulli trials. The PMF is given by

$$P\{X = k\} = \begin{cases} \binom{n}{k} p^k (1-p)^{n-k}, & 0 \le k \le n \\ 0, & \text{otherwise} \end{cases}$$

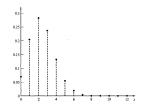


Figure: The PMF for the binomial random variable.

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#### Statement (Uniform Random Variable)

The Uniform random variable is a continuous random variable taking values between a and b with equal probabilities for intervals of equal length. The density function is given by

$$f_X(x) = egin{cases} rac{1}{b-a}, & a \leq x \leq b \ 0, & otherwise \end{cases}$$



Figure: The PDF for the uniform random variable.

#### Statement (Gaussian Random Variable)

The Gaussian, or normal, random variable  $\mathcal{N}(m, \sigma^2)$  is a continuous random variable described by the density function

$$f_X(x) = rac{1}{\sqrt{2\pi\sigma}} e^{-rac{(x-m)^2}{2\sigma^2}}$$

, where m,  $\sigma$ , and  $\sigma^2$  are named mean, standard deviation, and variance.

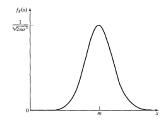


Figure: The PDF for the Gaussian random variable.

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# Statement (Q Function)

Assuming that X is a standard normal random variable  $\mathcal{N}(0,1)$ , the function Q(x) is defined as

$$Q(x) = P\{X > x\} = \int_{x}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{t^2}{2}} dt$$

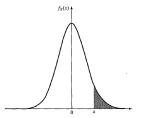


Figure: The Q-function as the area under the tail of a standard normal random variable.

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The Q function has the following properties,

• 
$$Q(-\infty) = 1$$
,  $Q(0) = 0.5$ ,  $Q(+\infty) = 0$ .  
•  $Q(-x) = 1 - Q(x)$ .

The important bounds on the Q function are

Q(x) 
$$\leq \frac{1}{2}e^{-\frac{x^2}{2}}, x \geq 0.$$
Q(x)  $< \frac{1}{\sqrt{2\pi x}}e^{-\frac{x^2}{2}}, x \geq 0.$ 
Q(x)  $< \frac{1}{\sqrt{2\pi x}}(1 - \frac{1}{x^2})e^{-\frac{x^2}{2}}, x > 1.$ 

For an  $\mathcal{N}(m, \sigma^2)$  random variable,

• 
$$F_X(x) = P\{X \le x\} = 1 - Q(\frac{x-m}{\sigma}).$$

# Important Random Variables

x	Q(x)	x	Q(x)	x	Q(x)
0.0	$5.000000  imes 10^{-01}$	2.4	$8.197534  imes 10^{-03}$	4.8	$7.933274  imes 10^{-07}$
0.1	$4.601722 \times 10^{-01}$	2.5	$6.209665  imes 10^{-03}$	4.9	$4.791830 \times 10^{-07}$
0.2	$4.207403  imes 10^{-01}$	2.6	$4.661189  imes 10^{-03}$	5.0	$2.866516 \times 10^{-07}$
0.3	$3.820886 \times 10^{-01}$	2.7	$3.466973  imes 10^{-03}$	5.1	$1.698268 \times 10^{-07}$
0.4	$3.445783 \times 10^{-01}$	2.8	$2.555131 \times 10^{-03}$	5.2	$9.964437  imes 10^{-06}$
0.5	$3.085375 \times 10^{-01}$	2.9	$1.865812 \times 10^{-03}$	5.3	$5.790128  imes 10^{-08}$
0.6	$2.742531 \times 10^{-01}$	3.0	$1.349898  imes 10^{-03}$	5.4	$3.332043 \times 10^{-08}$
0.7	$2.419637 \times 10^{-01}$	3.1	$9.676035  imes 10^{-04}$	5.5	$1.898956 \times 10^{-08}$
0.8	$2.118554 \times 10^{-01}$	3.2	$6.871378  imes 10^{-04}$	5.6	$1.071760 \times 10^{-08}$
0.9	$1.840601 \times 10^{-01}$	3.3	$4.834242  imes 10^{-04}$	5.7	$5.990378  imes 10^{-09}$
1.0	$1.586553  imes 10^{-01}$	3.4	$3.369291  imes 10^{-04}$	5.8	$3.315742  imes 10^{-09}$
1.1	$1.356661 \times 10^{-01}$	3.5	$2.326291  imes 10^{-04}$	5.9	$1.817507 \times 10^{-09}$
1.2	$1.150697 \times 10^{-01}$	3.6	$1.591086  imes 10^{-04}$	6.0	$9.865876 \times 10^{-10}$
1.3	$9.680049 \times 10^{-02}$	3.7	$1.077997  imes 10^{-04}$	6.1	$5.303426 \times 10^{-10}$
1.4	$8.075666 \times 10^{-02}$	3.8	$7.234806 \times 10^{-05}$	6.2	$2.823161 \times 10^{-10}$
1.5	$6.680720 \times 10^{-02}$	3.9	$4.809633  imes 10^{-05}$	6.3	$1.488226 \times 10^{-10}$
1.6	$5.479929  imes 10^{-02}$	4.0	$3.167124 \times 10^{-05}$	6.4	$7.768843  imes 10^{-11}$
1.7	$4.456546 \times 10^{-02}$	4.1	$2.065752 \times 10^{-05}$	6.5	$4.016001 \times 10^{-11}$
1.8	$3.593032 \times 10^{-02}$	4.2	$1.334576  imes 10^{-05}$	6.6	$2.055790 \times 10^{-11}$
1.9	$2.871656 \times 10^{-02}$	4.3	$8.539898  imes 10^{-06}$	6.7	$1.042099 \times 10^{-11}$
2.0	$2.275013 \times 10^{-02}$	4.4	$5.412542  imes 10^{-06}$	6.8	$5.230951 \times 10^{-12}$
2.1	$1.786442 \times 10^{-02}$	4.5	$3.397673  imes 10^{-06}$	6.9	$2.600125 \times 10^{-12}$
2.2	$1.390345 \times 10^{-02}$	4.6	$2.112456 \times 10^{-06}$	7.0	$1.279813 \times 10^{-12}$
2.3	$1.072411 \times 10^{-02}$	4.7	$1.300809 \times 10^{-06}$		

Table: Table of the Q Function,  $\Box \rightarrow \langle \Box \rangle \rightarrow \langle \Box \rangle$ 

# Example (Q Function)

X is a Gaussian random variable with mean 1 and variance 4. Therefore,

$$P(5 < X < 7) = F_X(7) - F_X(5)$$
  
= 1 - Q( $\frac{7-1}{2}$ ) - [1 - Q( $\frac{5-1}{2}$ )]  
= Q(2) - Q(3) \approx 0.0214

Statement (Functions of a Random Variable)

The CDF of the random variable Y = g(X) is

$$F_Y(y) = P\{\omega \in \Omega : g(X(\omega)) \le y\}$$

. In the special case that, for all y, the equation g(x) = y has a countable number of solutions  $\{x_i\}$ , and for all these solutions,  $g'(x_i)$  exists and is nonzero,

$$f_Y(y) = \sum_i \frac{f_X(x_i)}{|g'(x_i)|}$$

#### Example (Linear function of a normal variable)

if X is  $\mathcal{N}(m, \sigma^2)$ , then Y = aX + b is also a Gaussian random variable of the form  $\mathcal{N}(am + b, a^2\sigma^2)$ .

If y = ax + b = g(x), then x = (y - b)/a and g'(x) = a. So,

$$f_{Y}(y) = \frac{f_{X}(x)}{|g'(x)|}\Big|_{x=(y-b)/a}$$
  
=  $\frac{1}{a} \frac{1}{\sqrt{2\pi\sigma}} e^{-\frac{(x-m)^{2}}{2\sigma^{2}}}\Big|_{x=(y-b)/a}$   
=  $\frac{1}{\sqrt{2\pi}a\sigma} e^{-\frac{(y-b-m)^{2}}{2\sigma^{2}}}$   
=  $\frac{1}{\sqrt{2\pi}a\sigma} e^{-\frac{(y-b-m)^{2}}{2\sigma^{2}\sigma^{2}}}$ 

## Definition (Mean of Function)

The mean, expected value, or expectation of the random variable Y = g(X) is defined as

$$E\{g(X)\} = \int_{-\infty}^{\infty} g(x) f_X(x) dx$$

#### Definition (Mean of Function)

The mean, expected value, or expectation of the discrete random variable Y = g(X) is defined as

$$E\{g(X)\} = \sum_{i} g(x_i) P\{X = x_i\}$$

# Statistical Averages

#### Definition (Mean)

The mean, expected value, or expectation of the random variable X is defined as

$$E\{X\} = m_X = \int_{-\infty}^{\infty} x f_X(x) dx$$

### Definition (Mean)

The mean, expected value, or expectation of the discrete random variable X is defined as

$$E\{X\} = m_X = \sum_i x_i P\{X = x_i\}$$

- E(cX) = cE(X).
- 2 E(X + c) = c + E(X).
- E(c) = c.

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# Definition (Variance)

The variance of the random variable X is defined as

$$\sigma_X^2 = V(X) = E\{(X - E\{X\})^2\} = E\{X^2\} - (E\{X\})^2$$

- $V(cX) = c^2 V(X).$ **2** V(X + c) = V(X).
- **3** V(c) = 0.

Example (Bernoulli random variable)

If X is a Bernoulli random variable, E(X) = p and V(X) = p(1-p).

Example (Binomial random variable)

If X is a Binomial random variable, E(X) = np and V(X) = np(1-p).

Example (Uniform random variable)

If X is a Uniform random variable,  $E(X) = \frac{a+b}{2}$  and  $V(X) = \frac{(b-a)^2}{12}$ .

Example (Gaussian random variable)

If X is a Gaussian random variable, E(X) = m and  $V(X) = \sigma^2$ .

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#### Definition (Joint CDF)

Let X and Y represent two random variables. For these two random variables, the joint CDF is defined as

$$F_{X,Y}(x,y) = P(X \le x, Y \le y)$$

$$F_X(x) = F_{X,Y}(x,\infty).$$

 $P_Y(x) = F_{X,Y}(\infty, y).$ 

**3** If X and Y are statistically independent,  $F_{X,Y}(x,y) = F_X(x)F_Y(y)$ .

# Definition (Joint PDF)

Let X and Y represent two random variables. For these two random variables, the joint PDF is defined as

$$f_{X,Y}(x,y) = \frac{\partial^2 F_{X,Y}(x,y)}{\partial x \partial y}$$

• 
$$f_X(x) = \int_{-\infty}^{\infty} f_{X,Y}(x,y) dy.$$
  
•  $f_Y(y) = \int_{-\infty}^{\infty} f_{X,Y}(x,y) dx.$   
•  $\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f_{X,Y}(x,y) dx dy = 1.$   
•  $P\{(x,y) \in A\} = \int \int_{(x,y) \in A} f_{X,Y}(x,y) dx dy.$   
•  $F_{X,Y}(x,y) = \int_{-\infty}^{x} \int_{-\infty}^{y} f_{X,Y}(u,v) du dv.$   
• If X and Y are statistically independent,  $f_{X,Y}(x,y) = f_X(x) f_Y(y).$ 

## Definition (Conditional PDF)

The conditional PDF of the random variable Y, given that the value of the random variable X is equal to x, is defined as

$$f_{Y|X}(y|x) = \begin{cases} \frac{f_{X,Y}(x,y)}{f_X(x)}, & f_X(x) \neq 0\\ 0, & f_X(x) = 0 \end{cases}$$

# Definition (Mean)

The expected value of g(X, Y) is defined as  $E\{g(X, Y)\} = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} g(x, y) f_{X,Y}(x, y) dx dy$ 

#### Definition (Correlation)

R(X, Y) = E(XY) is called the correlation f X and Y.

#### Definition (Covariance)

The covariance of X and Y is defined as C(X, Y) = E(XY) - E(X)E(Y).

#### Definition (Correlation Coefficient)

The correlation coefficient of X and Y is defined as  $\rho_{X,Y} = C(X,Y)/(\sigma_X \sigma_Y)$ .

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- If  $\rho_{X,Y} = C(X, Y) = 0$ . i.e., E(XY) = E(X)E(Y), then X and Y are called uncorrelated.
- If X and Y are independent, E(XY) = E(X)E(Y), i.e., X and Y are uncorrelated.
- **3**  $|\rho_{X,Y}| \le 1.$
- If  $\rho_{X,Y} = 1$ , then Y = aX + b, where *a* is a positive.
- **5** If  $\rho_{X,Y} = -1$ , then Y = aX + b, where *a* is a negative.

#### Example (Moment calculation)

Assume that  $X \sim \mathcal{N}(3,4)$  and  $Y \sim \mathcal{N}(-1,2)$  are independent. If Z = X - Y and W = 2X + 3Y, then

$$E(Z) = E(X) - E(Y) = 3 + 1 = 4$$
  

$$E(W) = 2E(X) + 3E(Y) = 6 - 3 = 3$$
  

$$E(X^{2}) = V(X) + (E(X))^{2} = 4 + 9 = 13$$
  

$$E(Y^{2}) = V(Y) + (E(Y))^{2} = 2 + 1 = 3$$
  

$$E(XY) = E(X)E(Y) = -3$$
  

$$C(W, Z) = E(WZ) - E(W)E(Z) = E(2X^{2} - 3Y^{2} + XY) - 12 = 2$$

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### Statement (Multiple Functions of Multiple Random Variables)

If Z = g(X, Y) and W = h(X, Y) and the set of equations

$$\begin{cases} g(x,y) = z \\ h(x,y) = w \end{cases}$$

has a countable number of solutions  $\{(x_i, y_i)\}$ , and if at these points the determinant of the Jacobian matrix

$$J(x,y) = \begin{bmatrix} \frac{\partial z}{\partial x} & \frac{\partial z}{\partial y} \\ \frac{\partial w}{\partial x} & \frac{\partial w}{\partial y} \end{bmatrix}$$

is nonzero, then

$$f_{Z,W}(z,w) = \sum_{i} \frac{f_{X,Y}(x_i, y_i)}{|detJ(x_i, y_i)|}$$

#### Example (Magnitude and phase of two i.i.d Gaussian variables)

If X and Y are independent and identically distributed zero-mean Gaussian random variables with the variance  $\sigma^2$ , i.e.,  $X \sim \mathcal{N}(0, \sigma^2) \perp Y \sim \mathcal{N}(0, \sigma^2)$ , then the random variables  $V = \sqrt{X^2 + Y^2}$  and  $\Theta = \arctan \frac{Y}{X}$  are independent and have Rayleigh and uniform distribution, respectively, i.e.,  $V = \sqrt{X^2 + Y^2} \sim \mathcal{R}(\sigma) \perp \Theta = \arctan \frac{Y}{X} \sim \mathcal{U}[0, 2\pi]$ .

$$V=\sqrt{X^2+Y^2}$$
 and  $\Theta=\arctanrac{Y}{X}$  and

$$f_{X,Y}(x,y) = f_X(x)f_Y(y) = \frac{1}{2\pi\sigma^2}e^{-\frac{x^2+y^2}{2\sigma^2}}$$

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Example (Magnitude and phase of two i.i.d Gaussian variables)

$$J(x,y) = \begin{bmatrix} \frac{x}{\sqrt{x^2 + y^2}} & \frac{y}{\sqrt{x^2 + y^2}} \\ -\frac{y}{x^2 + y^2} & \frac{x}{x^2 + y^2} \end{bmatrix} \Rightarrow |\det J(x,y)| = \frac{1}{\sqrt{x^2 + y^2}} = \frac{1}{v}$$
$$\begin{cases} \sqrt{x^2 + y^2} = v \\ \arctan \frac{y}{x} = \theta \end{cases} \Rightarrow \begin{cases} x = v \cos \theta \\ y = v \sin \theta \end{cases}$$
$$f_{V,\Theta}(v,\theta) = v f_{X,Y}(v \cos \theta, v \sin \theta) = \frac{v}{2\pi\sigma^2} e^{-\frac{v^2}{2\sigma^2}} \end{cases}$$

Example (Magnitude and phase of two i.i.d Gaussian variables)

$$f_{\Theta}(\theta) = \int_{-\infty}^{\infty} f_{V,\Theta}(v,\theta) dv = \frac{1}{2\pi}, 0 \le \theta \le 2\pi$$
$$f_{V}(v) = \int_{-\infty}^{\infty} f_{V,\Theta}(v,\theta) d\theta = \frac{v}{\sigma^{2}} e^{-\frac{v^{2}}{2\sigma^{2}}}, v \ge 0$$

The magnitude and the phase are independent random variables since

$$f_{V,\Theta}(v,\theta) = f_{\Theta}(\theta)f_V(v)$$

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#### Statement (Jointly Gaussian Random Variables)

Jointly Gaussian random variables X and Y are distributed according to a joint PDF of the form

$$f_{X,Y}(x,y) = \frac{1}{2\pi\sigma_1\sigma_2\sqrt{1-\rho^2}} \\ \times \exp\left\{-\frac{1}{2(1-\rho^2)}\left[\frac{(x-m_1)^2}{\sigma_1^2} + \frac{(y-m_2)^2}{\sigma_2^2} - \frac{2\rho(x-m_1)(y-m_2)}{\sigma_1\sigma_2}\right]\right\}$$

✓ Two uncorrelated jointly Gaussian random variables are independent. Therefore, for jointly Gaussian random variables, independence and uncorrelatedness are equivalent.

## Definition (Multi-variate CDF)

Let  $\boldsymbol{X} = (X_1, \cdots, X_n)^T$  represent *n* random variables. For these random vector , the CDF is defined as

$$F_{\boldsymbol{X}}(\boldsymbol{x}) = F_{X_1,\cdots,X_n}(x_1,\cdots,x_n) = P(X_1 \leq x_1,\cdots,X_n \leq x_n)$$

#### Definition (Multi-variate PDF)

Let  $\boldsymbol{X} = (X_1, \cdots, X_n)^T$  represent *n* random variables. For these random vector , the PDF is defined as

$$f_{\mathbf{X}}(\mathbf{x}) = f_{X_1, \cdots, X_n}(x_1, \cdots, x_n) = \frac{\partial^n F_{X_1, \cdots, X_n}(x_1, \cdots, x_n)}{\partial x_1 \cdots \partial x_n}$$

## Definition (Joint Multi-variate CDF)

Let  $\mathbf{X} = (X_1, \dots, X_n)^T$  and  $\mathbf{Y} = (Y_1, \dots, Y_m)^T$  represent two random vectors. For these random vector, the joint CDF is defined as

$$F_{\boldsymbol{X},\boldsymbol{Y}}(\boldsymbol{x},\boldsymbol{y}) = P(X_1 \leq x_1, \cdots, X_n \leq x_n, Y_1 \leq y_1, \cdots, Y_m \leq y_m)$$

## Definition (Joint Multi-variate PDF)

Let  $\mathbf{X} = (X_1, \dots, X_n)^T$  and  $\mathbf{Y} = (Y_1, \dots, Y_m)^T$  represent two random vectors. For these random vector, the joint PDF is defined as

$$f_{\mathbf{X},\mathbf{Y}}(\mathbf{x},\mathbf{y}) = \frac{\partial^{n+m} F_{\mathbf{X},\mathbf{Y}}(\mathbf{x},\mathbf{y})}{\partial x_1 \cdots \partial x_n \partial y_1 \cdots \partial y_m}$$

## Definition (Mean)

The expected value of  $\boldsymbol{X}$  is defined as  $E(\boldsymbol{X}) = (E\{X_1\}, \cdots, E\{X_n\})$ 

## Definition (Correlation)

 $R(X, Y) = E(XY^{T})$  is called the correlation matrix of X and Y.

#### Definition (Covariance)

The covariance of X and Y is defined as  $C(\mathbf{X}, \mathbf{Y}) = E((\mathbf{X} - E(\mathbf{X}))(\mathbf{Y} - E(\mathbf{Y}))^T) = E(\mathbf{X}\mathbf{Y}^T) - E(\mathbf{X})E(\mathbf{Y})^T$ .

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- If  $f_{\mathbf{X}}(\mathbf{x}) = f_{X_1}(x_1) \cdots f_{X_n}(x_n)$ , then **X** is called mutually independent.
- **2** If C(X, X) is a diagonal matrix, then X is called mutually uncorrelated.
- If X is independent, then, X is uncorrelated.
- If  $f_{X,Y}(x, y) = f_X(x)f_Y(y)$ , then X and Y are called independent.
- **5** If C(X, Y) = 0, then X and Y are called uncorrelated.
- **(4)** If **X** and **Y** are independent, **X** and **Y** are uncorrelated.

#### Statement (Jointly Gaussian Random Variables)

Jointly Gaussian random variables  $\mathbf{X} = (X_1, \dots, X_n)^T$  are distributed according to a joint PDF of the form

$$f_{\boldsymbol{X}}(\boldsymbol{x}) = (2\pi |\boldsymbol{\Sigma}|)^{-\frac{n}{2}} \exp\left[\frac{-1}{2}(\boldsymbol{x} - \boldsymbol{m})^T \boldsymbol{\Sigma}^{-1}(\boldsymbol{x} - \boldsymbol{m})\right]$$

, where  $\mathbf{m} = E(\mathbf{X})$  and  $\mathbf{\Sigma} = C(\mathbf{X}, \mathbf{X})$  are the mean vector and covariance matrix and  $|\mathbf{\Sigma}|$  is the determinant of  $\mathbf{\Sigma}$ .

✓ Uncorrelated jointly Gaussian random variables are independent. Therefore, for jointly Gaussian random variables, independence and uncorrelatedness are equivalent.

## Theorem (Central Limit Theorem)

If  $\{X_i\}_{i=1}^n$  are n i.i.d. (independent and identically distributed) random variables, which each have the mean m and variance  $\sigma^2$ , then  $Y = \frac{1}{n} \sum_{i=1}^n X_i$  converges to  $\mathcal{N}(m, \frac{\sigma^2}{n})$ .

✓ The central limit theorem states that the sum of many i.i.d. random variables converges to a Gaussian random variable.

# Random Processes

Mohammad Hadi

Communication systems

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 $\checkmark$  A random process is a set of possible realizations of signal waveforms.

# Random Processes

## Example (Sample random process)

 $X(t) = A\cos(2\pi f_0 t + \Theta), \quad \Theta \sim U[0, 2\pi].$ 

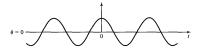






Figure: Sample functions of the example random process.

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# Random Processes

## Example (Sample random process)

 $X(t) = X, \quad X \sim U[-1,1].$ 

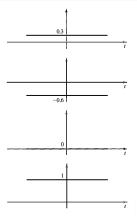


Figure: Sample functions of the example random process.

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✓ A random process is denoted by x(t; ω), where ω ∈ Ω is a random variable.

✓ For each  $\omega_i$ , there exists a deterministic time function  $x(t; \omega_i)$ , which is called a sample function or a realization.

✓ For the different outcomes at a fixed time  $t_0$ , the numbers  $x(t_0; \omega)$  constitute a random variable denoted by  $X(t_0)$ .

✓ At each time instant  $t_0$  and for each  $ω_i ∈ Ω$ , we have the number  $x(t_0; ω_i)$ .

#### Example (Sample random process)

Let  $\Omega = \{1, 2, 3, 4, 5, 6\}$  denote the sample space corresponding to the random experiment of throwing a die. For all  $\omega \in \Omega$ , let  $x(t; \omega) = \omega e^{-t} u(t)$  denote a random process. Then X(1) is a random variable taking values  $\{e^{-1}, 2e^{-1}, 3e^{-1}, 4e^{-1}, 5e^{-1}, 6e^{-1}\}$  and each has probability 1/6.

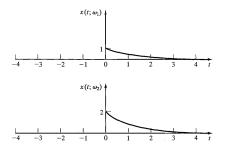


Figure: Sample functions of a random process.

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#### Definition (Mean Function)

The mean, or expectation, of the random process X(t) is a deterministic function of time denoted by  $m_X(t)$  that at each time instant to equals the mean of the random variable  $X(t_0)$ . That is,  $m_X(t) = E[X(t)] = \int_{-\infty}^{\infty} x f_{X(t)}(x) dx, \forall t$ .

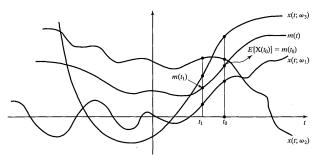


Figure: The mean of a random process.

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## Definition (Autocorrelation Function)

The autocorrelation function of the random process X(t) is defined as

$$R_X(t_1, t_2) = E[X(t_1)X(t_2)] = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} x_1 x_2 f_{X(t_1), X(t_2)}(x_1, x_2) dx_1 dx_2$$

# Statistical Averages

#### Example (Statistical averages)

If  $X(t) = A\cos(2\pi f_0 t + \Theta)$ ,  $\Theta \sim U[0, 2\pi]$ , then  $m_X(t) = 0$  and  $R_X(t_1, t_2) = \frac{A^2}{2}\cos(2\pi f_0(t_1 - t_2))$ .

$$m_{X}(t) = E[X(t)] = E[A\cos(2\pi f_{0}t + \Theta)] = \int_{0}^{2\pi} A\cos(2\pi f_{0}t + \theta) \frac{1}{2\pi} d\theta = 0$$

$$\begin{aligned} R_X(t_1, t_2) &= E[X(t_1)X(t_2)] \\ &= E[A\cos(2\pi f_0 t_1 + \Theta)A\cos(2\pi f_0 t_2 + \Theta)] \\ &= E[\frac{A^2}{2}\cos(2\pi f_0(t_1 - t_2)) + \frac{A^2}{2}\cos(2\pi f_0(t_1 + t_2) + 2\Theta)] \\ &= \frac{A^2}{2}\cos(2\pi f_0(t_1 - t_2)) \end{aligned}$$

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## Example (Statistical averages)

If 
$$X(t) = X$$
,  $X \sim U[-1, 1]$ , then  $m_X(t) = 0$  and  $R_X(t_1, t_2) = \frac{1}{3}$ .

$$m_{X}(t) = E[X(t)] = E[X] = \frac{-1+1}{2} = 0$$
$$R_{X}(t_{1}, t_{2}) = E[X^{2}] = \frac{(1-(-1))^{2}}{12} = \frac{1}{3}$$

## Definition (Wide-Sense Stationary (WSS))

A process X(t) is WSS if the following conditions are satisfied

- $m_x(t) = E[X(t)]$  is independent of t.
- $R_X(t_1, t_2)$  depends only on the time difference  $\tau = t_1 t_2$  and not on  $t_1$  and  $t_2$  individually.
- $R_X(t_1, t_2) = R_X(t_2, t_1).$
- 2 If X(t) is WSS,  $R_X(\tau) = R_X(-\tau)$ .

## Example (WSS)

If  $X(t) = A\cos(2\pi f_0 t + \Theta)$ ,  $\Theta \sim U[0, 2\pi]$ , then  $m_X(t) = 0$  and  $R_X(t_1, t_2) = \frac{A^2}{2}\cos(2\pi f_0(t_1 - t_2))$  and therefore, X(t) is WSS.

$$m_{x}(t) = E[A\cos(2\pi f_{0}t + \Theta)] = \int_{0}^{2\pi} A\cos(2\pi f_{0}t + \theta)\frac{1}{2\pi}d\theta = 0$$

$$\begin{aligned} R_X(t_1, t_2) &= E[A\cos(2\pi f_0 t_1 + \Theta)A\cos(2\pi f_0 t_2 + \Theta)] \\ &= E[\frac{A^2}{2}\cos(2\pi f_0(t_1 - t_2)) + \frac{A^2}{2}\cos(2\pi f_0(t_1 + t_2) + 2\Theta)] \\ &= \frac{A^2}{2}\cos(2\pi f_0(t_1 - t_2)) \end{aligned}$$

## Example (WSS)

If  $X(t) = A\cos(2\pi f_0 t + \Theta)$ ,  $\Theta \sim U[0,\pi]$ , then  $m_X(t) = -2\frac{A}{\pi}\sin(2\pi f_0 t)$ and  $R_X(t_1,t_2) = \frac{A^2}{2}\cos(2\pi f_0(t_1-t_2))$  and therefore, X(t) is not WSS.

$$m_{x}(t) = E[A\cos(2\pi f_{0}t + \Theta)] = \int_{0}^{\pi} A\cos(2\pi f_{0}t + \theta)\frac{1}{\pi}d\theta = -2\frac{A}{\pi}\sin(2\pi f_{0}t)$$

$$\begin{aligned} R_X(t_1, t_2) &= E[A\cos(2\pi f_0 t_1 + \Theta)A\cos(2\pi f_0 t_2 + \Theta)] \\ &= E[\frac{A^2}{2}\cos(2\pi f_0(t_1 - t_2)) + \frac{A^2}{2}\cos(2\pi f_0(t_1 + t_2) + 2\Theta)] \\ &= \frac{A^2}{2}\cos(2\pi f_0(t_1 - t_2)) \end{aligned}$$

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#### Definition (Independent Processes)

Two random processes X(t) and Y(t) are independent if for all positive integers m, n, and for all  $t_1, t_2, \dots, t_n$  and  $\tau_1, \tau_2, \dots, \tau_m$  the random vectors  $(X(t_1), X(t_2), \dots, X(t_n))$  and  $(Y(\tau_1), Y(\tau_2), \dots, Y(\tau_m))$  are independent.

#### Definition (Uncorrelated Processes)

Two random processes X(t) and Y(t) are uncorrelated if for all positive integers m, n, and for all  $t_1, t_2, \dots, t_n$  and  $\tau_1, \tau_2, \dots, \tau_m$  the random vectors  $(X(t_1), X(t_2), \dots, X(t_n))$  and  $(Y(\tau_1), Y(\tau_2), \dots, Y(\tau_m))$  are uncorrelated.

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- The independence of random processes implies that they are uncorrelated.
- Interpretated provide the second s
- For the important class of Gaussian processes, the independence and uncorrelatedness are equivalent.

## Definition (Cross Correlation)

The cross correlation between two random processes X(t) and Y(t) is defined as  $R_{XY}(t_1, t_2) = E[X(t_1)Y(t_2)]$ .

## Definition (Jointly WSS)

Two random processes X(t) and Y(t) are jointly wide-sense stationary, or simply jointly stationary, if both X(t) and Y(t) are individually stationary and the cross-correlation  $R_{XY}(t_1, t_2)$  depends only on  $\tau = t_1 - t_2$ .

$$R_{XY}(t_1,t_2) = R_{YX}(t_2,t_1).$$

**2** For jointly WSS random processes X(t) and Y(t),  $R_{XY}(\tau) = R_{YX}(-\tau)$ .

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#### Example (Jointly WSS)

Assuming that the two random processes X(t) and Y(t) are jointly stationary, determine the autocorrelation of the process Z(t) = X(t) + Y(t).

$$R_{Z}(t + \tau, t) = E[Z(t + \tau)Z(t)]$$
  
=  $E[(X(t + \tau) + Y(t + \tau))(X(t) + Y(t))]$   
=  $R_{X}(\tau) + R_{Y}(\tau) + R_{XY}(\tau) + R_{XY}(-\tau)$ 

If a stationary process X(t) with mean  $m_x$  and autocorrelation function  $R_X(\tau)$  is passed through an LTI system with impulse response h(t), the input and output processes X(t) and Y(t) will be jointly stationary with

$$m_Y = m_X \int_{-\infty}^{\infty} h(t) dt$$

$$R_{XY}(\tau) = R_X(\tau) * h(-\tau)$$
$$R_Y(\tau) = R_{XY}(\tau) * h(\tau) = R_X(\tau) * h(\tau) * h(-\tau)$$

$$X(t)$$
  $h(t)$   $Y(t)$ 

Figure: A random process passing through an LTI system.

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If a stationary process X(t) with mean  $m_x$  and autocorrelation function  $R_X(\tau)$  is passed through an LTI system with impulse response h(t), the input and output processes X(t) and Y(t) will be jointly stationary.

$$E[Y(t)] = E\left[\int_{-\infty}^{\infty} X(\tau)h(t-\tau)d\tau\right]$$
$$= \int_{-\infty}^{\infty} E[X(\tau)]h(t-\tau)d\tau]$$
$$= \int_{-\infty}^{\infty} m_X h(t-\tau)d\tau$$
$$= m_X \int_{-\infty}^{\infty} h(u)du = m_Y$$

If a stationary process X(t) with mean  $m_x$  and autocorrelation function  $R_X(\tau)$  is passed through an LTI system with impulse response h(t), the input and output processes X(t) and Y(t) will be jointly stationary.

$$\begin{split} E[X(t_1)Y(t_2)] &= E[X(t_1)\int_{-\infty}^{\infty}X(s)h(t_2-s)ds] \\ &= \int_{-\infty}^{\infty}E[X(t_1)X(s)]h(t_2-s)ds \\ &= \int_{-\infty}^{\infty}R_X(t_1-s)h(t_2-s)ds \\ &= \int_{-\infty}^{\infty}R_X(t_1-t_2-u)h(-u)du = R_X(\tau)*h(-\tau) = R_{XY}(\tau) \end{split}$$

If a stationary process X(t) with mean  $m_x$  and autocorrelation function  $R_X(\tau)$  is passed through an LTI system with impulse response h(t), the input and output processes X(t) and Y(t) will be jointly stationary.

$$\begin{split} E[Y(t_1)Y(t_2)] &= E[Y(t_2)\int_{-\infty}^{\infty} X(s)h(t_1 - s)ds] \\ &= \int_{-\infty}^{\infty} E[X(s)Y(t_2)]h(t_1 - s)ds \\ &= \int_{-\infty}^{\infty} R_{XY}(s - t_2)h(t_1 - s)ds \\ &= \int_{-\infty}^{\infty} R_{XY}(u)h(t_1 - t_2 - u)du = R_{XY}(\tau) * h(\tau) = R_Y(\tau) \end{split}$$

## Example (Differentiateor)

Assume a stationary process passes through a differentiator. What are the mean and autocorrelation functions of the output? What is the cross correlation between the input and output?

Since  $h(t) = \delta'(t)$ ,

$$m_Y = m_x \int_{-\infty}^{\infty} h(t) dt = m_x \int_{-\infty}^{\infty} \delta'(t) dt = 0$$

 $R_{XY}(\tau) = R_X(\tau) * h(-\tau) = R_X(\tau) * \delta'(-\tau) = -R_X(\tau) * \delta'(\tau) = -\frac{dR_X(\tau)}{d\tau}$ 

$$R_Y(\tau) = R_{XY}(\tau) * h(\tau) = -\frac{dR_X(\tau)}{d\tau} * \delta'(\tau) = -\frac{d^2R_X(\tau)}{d\tau^2}$$

## Example (Hilbert Transform)

Assume a stationary process passes through a Hilbert filter. What are the mean and autocorrelation functions of the output? What is the cross correlation between the input and output?

Assume that  $R_X(\tau)$  has no DC component. Since  $h(t) = 1/(\pi t)$ ,

$$m_Y = m_x \int_{-\infty}^{\infty} h(t) dt = m_x \int_{-\infty}^{\infty} \frac{1}{\pi t} dt = 0$$

$$R_{XY}(\tau) = R_X(\tau) * h(-\tau) = R_X(\tau) * \frac{-1}{\pi\tau} = -\widehat{R}_X(\tau)$$

$$R_Y(\tau) = R_{XY}(\tau) * h(\tau) = -\widehat{R}_X(\tau) * \frac{1}{\pi\tau} = -\widehat{R}_X(\tau) = R_X(\tau)$$

#### Definition (Truncated Fourier Transform)

The truncated Fourier transform of a realization of the random process  $X(t; \omega_i)$  over an interval [-T/2, T/2] is defined by

$$X_{T}(f;\omega_{i})=\int_{-T/2}^{T/2}x(t;\omega_{i})e^{-j2\pi ft}dt$$

#### Definition (Power Spectral Density)

The power spectral density of the random process X(t) is defined by

$$S_X(f) = \lim_{T \to \infty} \frac{1}{T} E[|X_T(f;\omega)|^2]$$

#### Theorem (Wiener-Khinchin)

For a stationary random process X(t), the power spectral density is the Fourier transform of the autocorrelation function, i.e.,

$$S_X(f) = \mathcal{F}[R_X(\tau)] = \int_{-\infty}^{\infty} R_X(\tau) e^{-j2\pi f \tau} d\tau$$

## Definition (Power)

The power in the random process X(t) is obtained by

$$P_X = \int_{-\infty}^{\infty} S_X(f) df = \mathcal{F}^{-1}[S_X(f)]|_{\tau=0} = R_X(0)$$

## Definition (Cross Power Spectral Density)

For the jointly stationary random processes X(t) and Y(t), the cross power spectral density is the Fourier transform of the cross correlation function, i.e.,

$$S_{XY}(f) = \mathcal{F}[R_{XY}(\tau)] = \int_{-\infty}^{\infty} R_{XY}(\tau) e^{-j2\pi f \tau} d\tau$$

#### Example (Wiener-Khinchin)

If  $X(t) = A\cos(2\pi f_0 t + \Theta)$ ,  $\Theta \sim U[0, 2\pi]$ , then  $R_X(\tau) = \frac{A^2}{2}\cos(2\pi f_0\tau)$ and therefore,  $S_X(f) = \frac{A^2}{4}[\delta(f - f_0) + \delta(f + f_0)]$  and  $P_X = \frac{A^2}{2}$ .

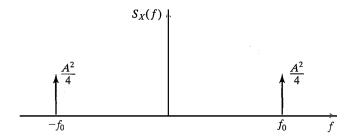


Figure: Power spectral density of the example random process.

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## Example (Wiener-Khinchin)

If 
$$X(t) = X$$
,  $X \sim U[-1,1]$ , then  $R_X(\tau) = \frac{1}{3}$  and therefore,  $S_X(f) = \frac{1}{3}\delta(f)$  and  $P_X = \frac{1}{3}$ .

# Statement (LTI System with Random Input)

If a stationary process X(t) with mean  $m_x$  and autocorrelation function  $R_X(\tau)$  is passed through an LTI system with impulse response h(t) and frequency response H(f), the input and output processes X(t) and Y(t) will be jointly stationary with

$$m_Y = m_x \int_{-\infty}^{\infty} h(t) dt \leftrightarrow m_y = m_x H(0)$$

$$R_{XY}(\tau) = R_X(\tau) * h(-\tau) \leftrightarrow S_{XY}(f) = H^*(f)S_X(f)$$
$$R_{YX}(\tau) = R_{XY}(-\tau) \leftrightarrow S_{YX}(f) = S^*_{XY}(f) = H(f)S_X(f)$$
$$R_Y(\tau) = R_{XY}(\tau) * h(\tau) = R_X(\tau) * h(\tau) * h(-\tau) \leftrightarrow S_Y(f) = |H(f)|^2 S_X(f)$$

### Statement (LTI System with Random Input)

If a stationary process X(t) with mean  $m_x$  and autocorrelation function  $R_X(\tau)$  is passed through an LTI system with impulse response h(t) and frequency response H(f), the input and output processes X(t) and Y(t) will be jointly stationary.

$$x(t)$$
  $h(t)$   $y(t)$ 

$$S_{X}(f) \longrightarrow H^{*}(f) \longrightarrow S_{XY}(f)$$

$$H(f) \longrightarrow S_{YX}(f)$$

$$|H(f)|^{2} \longrightarrow S_{Y}(f)$$

Figure: Input-output relations for the power spectral density and the cross-spectral density.

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### Example (Power spectral densities for a differentiator)

If  $X(t) = A\cos(2\pi f_0 t + \Theta)$ ,  $\Theta \sim U[0, 2\pi]$  passes through a differentiator, we have  $S_Y(f) = \pi^2 f_0^2 A^2 [\delta(f - f_0) + \delta(f + f_0)]$  and  $S_{XY}(f) = \frac{j\pi A^2 f_0}{2} [\delta(f + f_0) - \delta(f - f_0)]$ .

$$S_{Y}(f) = 4\pi^{2}f^{2}\frac{A^{2}}{4}[\delta(f-f_{0}) + \delta(f+f_{0})] = \pi^{2}f_{0}^{2}A^{2}[\delta(f-f_{0}) + \delta(f+f_{0})]$$

$$S_{XY}(f) = -j2\pi f \frac{A^2}{4} [\delta(f - f_0) + \delta(f + f_0)] = \frac{j\pi A^2 f_0}{2} [\delta(f + f_0) - \delta(f - f_0)]$$

### Example (Power spectral densities for a differentiator)

If X(t) = X,  $X \sim U[-1,1]$  passes through a differentiator, we have  $S_Y(f) = S_{XY}(f) = 0$ .

$$S_Y(f) = 4\pi^2 f^2 \frac{1}{3}\delta(f) = 0$$
$$S_{XY}(f) = -j2\pi f \frac{1}{3}\delta(f) = 0$$

#### Example (Power Spectral Density of a Sum Process)

Let Z(t) = X(t) + Y(t), where X(t) and Y(t) are jointly stationary random processes. Also assume that X(t) and Y(t) are uncorrelated and at least one of them has zero mean. Then,  $S_Z(f) = S_X(f) + S_Y(f)$ .

Since 
$$R_{XY}(\tau) = m_X m_Y = 0$$
,  
 $R_Z(\tau) = R_X(\tau) + R_Y(\tau) + R_{XY}(\tau) + R_{XY}(-\tau) = R_X(\tau) + R_Y(\tau)$ . So,  
 $S_Z(f) = \mathcal{F}\{R_Z(\tau)\} = S_X(f) + S_Y(f)$ 

# Gaussian, White, and Bandpass Processes

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Communication systems

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## Definition (Gaussian Random Process)

A random process X(t) is a Gaussian process if for all n and all  $(t_1, t_2, \dots, t_n)$ , the random variables  $\{X(t_i)\}_{i=1}^n$  have a jointly Gaussian density function.

For a Gassian random process,

- At any time instant  $t_0$ , the random variable  $X(t_0)$  is Gaussian.
- 2 At any two points  $t_1$ ,  $t_2$ , random variables  $(X(t_1), X(t_2))$  are distributed according to a two-dimensional jointly Gaussian distribution.

## Example (Gaussian Random Process)

Let X(t) be a zero-mean stationary Gaussian random process with the power spectral density  $S_X(f) = 5 \sqcap (f/1000)$ . Then,  $X(3) \sim \mathcal{N}(0, 5000)$ .

$$m = m_{X(3)} = m_X = 0$$
  
$$\sigma^2 = V[X(3)] = E[X^2(3)] - (E[X(3)])^2 = E[X(3)X(3)] = R_X(0) = P_X$$
  
$$\sigma^2 = P_X = \int_{-\infty}^{\infty} S_X(f) df = 5000$$

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### Definition (Jointly Gaussian Random Processes)

The random processes X(t) and Y(t) are jointly Gaussian if for all n, m and all  $(t_1, t_2, \dots, t_n)$  and  $(\tau_1, \tau_2, \dots, \tau_m)$ , the random vector  $(X(t_1), X(t_2), \dots, X(t_n), Y(\tau_1), Y(\tau_2), \dots, Y(\tau_m))$  is distributed according to an n + m dimensional jointly Gaussian distribution.

For jointly Gassian random processes,

- If the Gaussian process X(t) is passed through an LTI system, then the output process Y(t) will also be a Gaussian process. Moreover, X(t) and Y(t) will be jointly Gaussian processes.
- For jointly Gaussian processes, uncorrelatednesss and independence are equivalent.

#### Example (Jointly Gaussian Random Processes)

Let X(t) be a zero-mean stationary Gaussian random process with the power spectral density  $S_X(f) = 5 \sqcap (f/1000)$ . If X(t) passes a differentiator, the output random process  $Y(3) \sim \mathcal{N}(0, 1.6 \times 10^{10})$ .

Since 
$$H(f) = 2\pi f$$
,  
 $m = m_{Y(3)} = m_X H(0) = 0$   
 $\sigma^2 = V[Y(3)] = E[Y^2(3)] - (E[Y(3)])^2 = E[Y(3)Y(3)] = R_Y(0) = P_Y$   
 $\sigma^2 = P_Y = \int_{-\infty}^{\infty} |H(f)|^2 S_X(f) df = 1.6 \times 10^{10}$ 

### Definition (White Random Process)

A random process X(t) is called a white process if it has a flat power spectral density, i.e., if  $S_X(f) = \frac{N_0}{2}$  equals the constant  $\frac{N_0}{2}$  for all f.

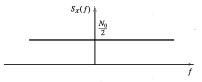


Figure: Power spectrum of a white process.

The power content of a white process

$$P_X = \int_{-\infty}^{\infty} S_X(f) df = \int_{-\infty}^{\infty} \frac{N_0}{2} df = \infty$$

A white process is not a meaningful physical process.

The autocorrelation function of a white process is

$$R_X(\tau) = \mathcal{F}^{-1}\{S_X(f)\} = \frac{N_0}{2}\delta(\tau)$$

- If we sample a zero-mean white process at two points  $t_1$  and  $t_2$  ( $t_1 \neq t_2$ ), the resulting random variables will be uncorrelated.
- If the zero-mean random process is white and also Gaussian, any pair of random variables  $X(t_1)$ ,  $X(t_2)$ , where  $t_1 \neq t_2$ , will also be independent.

#### Definition (Lowpass Random Process)

A WSS random process X(t) is called lowpass if its autocorrelation  $R_X(\tau)$  is a lowpass signal.

# Definition (Bandpass Random Process)

A zero-mean real WSS random process X(t) is called bandpass if its autocorrelation  $R_X(\tau)$  is a bandpass signal.

✓ For a bandpass process, the power spectral density is located around frequencies  $\pm f_c$ , and for lowpass processes, the power spectral density is located around zero frequency.

# Definition (In-phase/Quadrature Random Process)

The in-phase and quadrature components of a bandpass random process X(t) are defined as

$$\begin{aligned} X_c(t) &= X(t)\cos(2\pi f_c t) + \hat{X}(t)\sin(2\pi f_c t) \\ X_s(t) &= \hat{X}(t)\cos(2\pi f_c t) - X(t)\sin(2\pi f_c t) \end{aligned}$$

# Definition (Lowpass Equivalent Random Process)

The lowpass equivalent random process of a bandpass random process X(t) is defined as

$$X_l(t) = X_c(t) + jX_s(t)$$

# Theorem (In-phase/Quadrature Random Process)

For the in-phase and quadrature components of a bandpass random process X(t),

- **()**  $X_c(t)$  and  $X_s(t)$  are jointly WSS zero-mean random processes.
- 2  $X_c(t)$  and  $X_s(t)$  are both lowpass processes.

**③**  $X_c(t)$  and  $X_s(t)$  have the same power spectral density as

$$S_{X_c}(f) = S_{X_s}(f) = [S_X(f+f_c) + S_X(f-f_c)] \sqcap (\frac{f}{2f_c})$$

The cross-spectral density of the components are

$$S_{X_cX_s}(f) = -S_{X_sX_c}(f) = j[S_X(f+f_c) - S_X(f-f_c)] \sqcap (\frac{f}{2f_c})$$

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Theorem (Lowpass Equivalent Random Process)

For the lowpass equivalent of a bandpass random process X(t),

 $S_{X_l}(f) = 4S_X(f+f_c)u(f+f_c)$ 

$$S_X(f) = \frac{1}{4} [S_{X_l}(f - f_c) + S_{X_l}(-f - f_c)]$$

$$R_{X_l}(\tau) = 2(R_X(\tau) + j\widehat{R_X}(\tau))e^{-j2\pi f_c \tau}$$

## Example (In-phase autocorrelation)

The autocorrelation of the in-phase component of a bandpass random process X(t) is  $R_{X_c}(\tau) = R_X(\tau) \cos(2\pi f_c \tau) + \widehat{R_X}(\tau) \sin(2\pi f_c \tau)$ .

$$\begin{aligned} R_{X_c}(t+\tau,t) &= E\{X_c(t+\tau)X_c(t)\} \\ &= E\{[X(t+\tau)\cos(2\pi f_c(t+\tau)) + \hat{X}(t+\tau)\sin(2\pi f_c(t+\tau))] \\ &\times [X(t)\cos(2\pi f_c t) + \hat{X}(t)\sin(2\pi f_c t)]\} \\ &= R_X(\tau)\cos(2\pi f_c(t+\tau))\cos(2\pi f_c t) \\ &+ R_{\hat{X}\hat{X}}(t+\tau,t)\cos(2\pi f_c(t+\tau))\sin(2\pi f_c t) \\ &+ R_{\hat{X}\hat{X}}(t+\tau,t)\sin(2\pi f_c(t+\tau))\cos(2\pi f_c t) \\ &+ R_{\hat{X}\hat{X}}(t+\tau,t)\sin(2\pi f_c(t+\tau))\sin(2\pi f_c t) \\ &= R_X(\tau)\cos(2\pi f_c \tau) + \widehat{R_X}(\tau)\sin(2\pi f_c \tau) \end{aligned}$$

# Thermal and Filtered Noise

Mohammad Hadi

Communication systems

Fall 2020 91 / 102

 $\checkmark$  The thermal noise, which is produced by the random movement of electrons due to thermal agitation, is usually modeled by a white Gaussian process.

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# Statement (Thermal Noise)

Quantum mechanical analysis of the thermal noise shows that it has a power spectral density given by  $S_n(f) = 0.5hf/(e^{\frac{hf}{KT}} - 1)$ , which can be approximated by  $KT/2 = N_0/2$  for f < 2 THz, where  $h = 6.6 \times 10^{-34}$  J×sec denotes Planck's constan,  $K = 1.38 \times 10^{-23}$  J/K is Boltzmann's constant, and T denotes the temperature in degrees Kelvin. Further, the noise originates from many independent random particle movements.

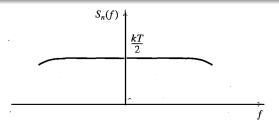


Figure: Power spectrum of thermal noise.

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#### Statement (Thermal Noise Model)

The thermal noise is assumed to have the following properties,

- Thermal noise is a stationary process.
- Intermal noise is a zero-mean process.
- Thermal noise is a Gaussian process.
- Thermal noise is a white process with a PSD  $S_n(f) = \frac{KT}{2} = \frac{N_0}{2}$ .

#### Statement (Filtered Noise Process)

The PSD of an ideally bandpass filtered noise is

$$S_X(f) = \frac{N_0}{2} |H(f)|^2$$

## Example (Filtered Noise Process)

If the Gaussian white noise passes through the shown filter, the PSD of the filtered noise is

$$S_X(f) = rac{N_0}{2} |H(f)|^2 = egin{cases} rac{N_0}{2}, & |f-f_c| \leq N \ 0, & ext{otherwise} \end{cases}$$

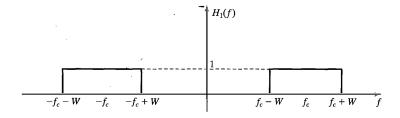


Figure: Filter transfer function H(f).

For a filtered white Gaussian noise, the following properties for  $X_c(t)$  and  $X_s(t)$  can be proved.

- $X_c(t)$  and  $X_s(t)$  are zero-mean, lowpass, jointly WSS, and jointly Gaussian random processes.
- If the power in process X(t) is  $P_X$ , then the power in each of the processes  $X_c(t)$  and  $X_s(t)$  is also  $P_x$ .
- Processes  $X_c(t)$  and  $X_s(t)$  have a common power spectral density, i.e.,  $S_{X_c}(f) = S_{X_s}(f) = [S_X(f + f_c) + S_X(f - f_c)] \sqcap (\frac{f}{2f_c}).$
- If  $f_c$  and  $-f_c$  are the axis of symmetry of the positive and negative frequencies, respectively, then  $X_c(t)$  and  $X_s(t)$  will be independent processes.

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#### Example (Filtered Noise Process)

For the bandpass white noise at the output of filter given below, power spectral density of the process  $Z(t) = aX_c(t) + bX_s(t)$  is  $S_Z(f) = N_0(a^2 + b^2) \sqcap (\frac{f}{2W}))$ .

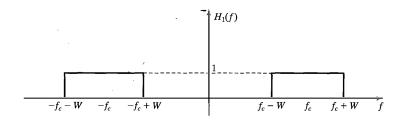


Figure: Filter transfer function H(f).

# Example (Filtered Noise Process (cont.))

For the bandpass white noise at the output of filter given below, power spectral density of the process  $Z(t) = aX_c(t) + bX_s(t)$  is  $S_Z(f) = N_0(a^2 + b^2) \sqcap (\frac{f}{2W})$ .

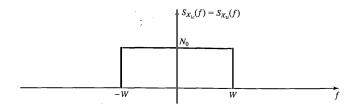


Figure: Power spectral densities of the in-phase and quadrature components of the example filtered noise.

## Example (Filtered Noise Process (cont.))

For the bandpass white noise at the output of filter given below, power spectral density of the process  $Z(t) = aX_c(t) + bX_s(t)$  is  $S_Z(f) = N_0(a^2 + b^2) \sqcap (\frac{f}{2W})$ .

Since  $f_c$  is the axis of symmetry of the noise power spectral density, the in-phase and quadrature components of the noise will be independent with zero mean. So,

$$R_{Z}(\tau) = E\{[aX_{c}(t+\tau)+bX_{s}(t+\tau)][aX_{c}(t)+bX_{s}(t)]\} = a^{2}R_{X_{c}}(\tau)+b^{2}R_{X_{s}}(\tau)$$

Since  $S_{X_c}(f) = S_{X_s}(f) = N_0 \sqcap (\frac{f}{2W})$ ,

$$S_Z(f) = a^2 S_{X_c}(f) + b^2 S_{X_s}(f) = N_0(a^2 + b^2) \sqcap (\frac{f}{2W})$$

# Definition (Noise Equivalent Bandwidth)

The noise equivalent bandwidth of a filter with the frequency response H(f) is defined as  $B_{neq} = \frac{\int_{-\infty}^{\infty} |H(f)|^2 df}{2H_{max}^2}$ , where  $H_{max}$  denotes the maximum of |H(f)| in the passband of the filter.

✓ The power content of the filtered noise is  $P_X = \int_{-\infty}^{\infty} |H(f)|^2 S_n(f) df$ =  $\frac{N_0}{2} \int_{-\infty}^{\infty} |H(f)|^2 df = N_0 B_{neq} H_{max}^2$ 

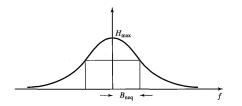


Figure: Noise equivalent bandwidth of a typical filter.

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# Example (Noise Equivalent Bandwidth)

The noise equivalent bandwidth of a lowpass RC filter is  $\frac{1}{4RC}$ .

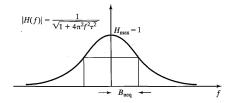


Figure: Frequency response of a lowpass f RC filter.

$$H(f) = \frac{1}{1 + j2\pi fRC} \Rightarrow |H(f)| = \frac{1}{\sqrt{1 + 4\pi^2 f^2 R^2 C^2}} \Rightarrow H_{max} = 1$$
$$B_{neq} = \frac{\int_{-\infty}^{\infty} |H(f)|^2 df}{2H_{max}^2} = \frac{\frac{1}{2RC}}{2} = \frac{1}{4RC}$$

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# The End

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