

# Natural Frequencies

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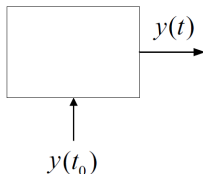
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# Natural Frequencies of a Network Variable

# Zero-input Response



**Figure:** Zero-input response for an LTI circuit. Natural frequencies of a variable are the constant coefficients appeared in the exponents of the zero-input response giving that variable. The number of natural frequencies of a variable is equal to or less than the circuit order.  $s_i$  is a frequency response of order  $n_i$ .

$$\sum_{k=0}^n a_k y^{(k)}(t) = 0, \quad y(0^-), y'(0^-), \dots, y^{(n-1)}(0^-)$$

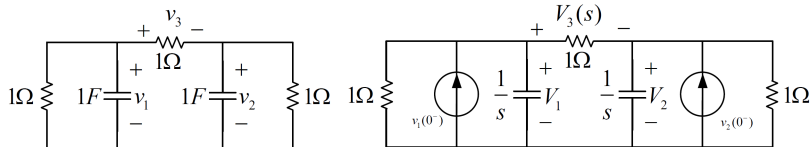
$$\sum_{k=0}^n [a_k s^k Y(s) - \sum_{k'=1}^k s^{k-k'} y^{(k'-1)}(0^-)] = 0 \Rightarrow Y(s) \sum_{k=0}^n a_k s^k - F_0(s) = 0 \Rightarrow Y(s) = \frac{F_0(s)}{\sum_{k=0}^n a_k s^k}$$

$$y(t) = \sum_{i=1}^r \sum_{j=1}^{n_i} K_{i,j} t^{j-1} e^{s_i t}$$

# Natural Frequencies of a Network Variable

## Example (Natural frequencies of a network variable)

The voltages  $V_1$  and  $V_2$  in the circuit below have two simple natural frequencies.



$$\begin{bmatrix} s+2 & -1 \\ -1 & s+2 \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} v_1(0^-) \\ v_2(0^-) \end{bmatrix} \Rightarrow \begin{cases} V_1 = \frac{(s+2)v_1(0^-) + v_2(0^-)}{(s+1)(s+3)} \\ V_2 = \frac{(s+2)v_2(0^-) + v_1(0^-)}{(s+1)(s+3)} \end{cases} \Rightarrow s = -1, -3$$

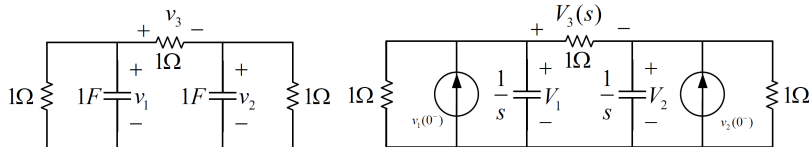
$$(v_1(0^-), v_2(0^-)) = (1, 0) \Rightarrow \begin{cases} V_1 = \frac{s+2}{(s+1)(s+3)} \\ V_2 = \frac{1}{(s+1)(s+3)} \end{cases}, \quad (v_1(0^-), v_2(0^-)) = (0, 0) \Rightarrow \begin{cases} V_1 = 0 \\ V_2 = 0 \end{cases}$$

$$(v_1(0^-), v_2(0^-)) = (1, -1) \Rightarrow \begin{cases} V_1 = \frac{1}{s+3} \\ V_2 = \frac{-1}{s+3} \end{cases}, \quad (v_1(0^-), v_2(0^-)) = (1, 1) \Rightarrow \begin{cases} V_1 = \frac{1}{s+1} \\ V_2 = \frac{1}{s+1} \end{cases}$$

# Natural Frequencies of a Network Variable

## Example (Natural frequencies of a network variable)

The voltages  $V_1$  and  $V_2$  in the circuit below have two simple natural frequency.

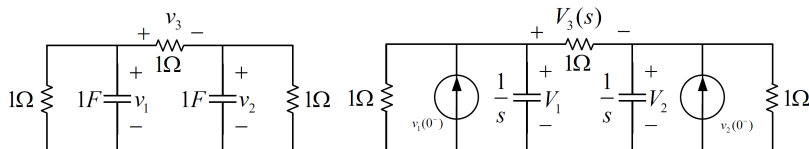


$$\begin{cases} v_1''(t) + 4v_1'(t) + 3v_1(t) = 0 \\ v_2''(t) + 4v_2'(t) + 3v_2(t) = 0 \end{cases} \Rightarrow s^2 + 4s + 3 = 0 \Rightarrow s = -1, -3$$

# Natural Frequencies of a Network Variable

## Example (Natural frequencies of a network variable)

The voltages  $V_3$  in the circuit below has one simple natural frequency.



$$\begin{bmatrix} s+2 & -1 \\ -1 & s+2 \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} v_1(0^-) \\ v_2(0^-) \end{bmatrix} \Rightarrow \begin{cases} V_1 = \frac{(s+2)v_1(0^-) + v_2(0^-)}{(s+1)(s+3)} \\ V_2 = \frac{(s+2)v_2(0^-) + v_1(0^-)}{(s+1)(s+3)} \end{cases}$$

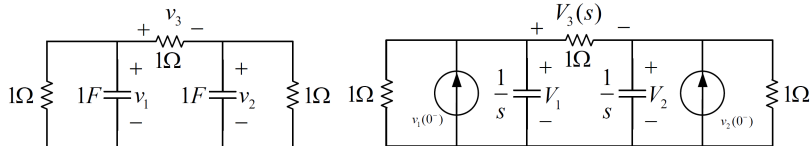
$$V_3 = V_1 - V_2 = \frac{(s+1)v_1(0^-) - (s+1)v_2(0^-)}{(s+1)(s+3)} = \frac{v_1(0^-) - v_2(0^-)}{s+3} \Rightarrow s = -3$$

$$(v_1(0^-), v_2(0^-)) = (0.5, -0.5) \Rightarrow V_3 = \frac{1}{s+3}, \quad (v_1(0^-), v_2(0^-)) = (1, 1) \Rightarrow V_3 = 0$$

# Natural Frequencies of a Network Variable

## Example (Natural frequencies of a network variable)

The voltages  $V_3$  in the circuit below has one simple natural frequencies.



$$\begin{cases} v_1''(t) + 4v_1'(t) + 3v_1(t) = 0 \\ v_2''(t) + 4v_2'(t) + 3v_2(t) = 0 \end{cases}, \quad v_3(t) = v_1(t) - v_2(t) \Rightarrow v_3''(t) + 4v_3'(t) + 3v_3(t) = 0 \Rightarrow s^2 + 4s + 3 = 0$$

$$v_3(t) = K_1 e^{-t} + K_2 e^{-3t}, \quad v_3(0^-) = v_1(0^-) - v_2(0^-), \quad v_3'(0^-) = -3[v_1(0^-) - v_2(0^-)]$$

$$\Rightarrow K_1 = 0, K_2 = v_1(0^-) - v_2(0^-) \Rightarrow v_3(t) = [v_1(0^-) - v_2(0^-)]e^{-3t}$$



# Types of Natural Frequencies

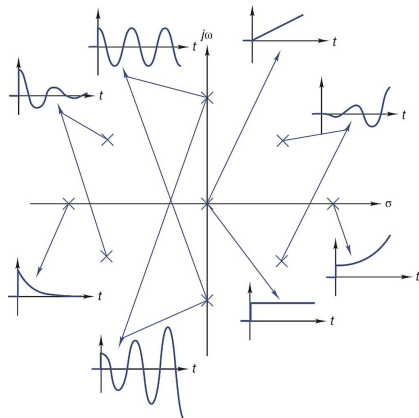


Figure: Natural frequencies can be simple or repeated, real, imaginary, or complex, and located in LHS, RHS, or  $j\omega$  axis of the complex plane.

# Types of Natural Frequencies

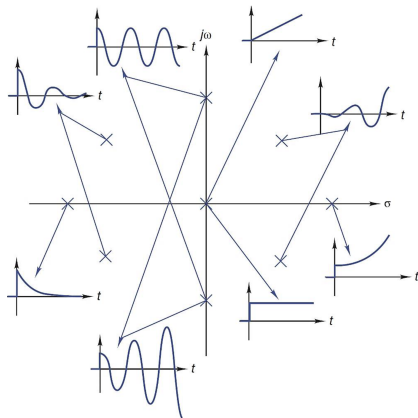


Figure: Types of natural frequencies.

- **Transfer function denominator factors:**  $s, (s + \alpha), (s + \alpha)^m, (s^2 + \omega^2), (s^2 + \omega^2)^m, (s + \alpha)^2 + \omega^2, [(s + \alpha)^2 + \omega^2]^m$
- **Time-domain behavior:**  $1, e^{-\alpha t}, t^m e^{-\alpha t}, \cos(\omega t + \theta), t^m \cos(\omega t + \theta), e^{-\alpha t} \cos(\omega t + \theta), t^m e^{-\alpha t} \cos(\omega t + \theta)$

# Types of Natural Frequencies

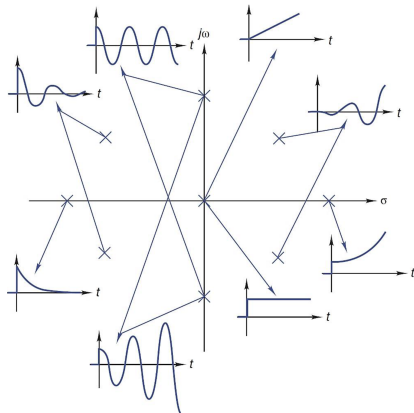


Figure: Impact of Natural frequencies on zero-input response stability.

- Stable circuit (strictly passive circuit):  $\Re\{s_i\} < 0, \forall i$
- Marginally stable circuit (passive circuit):  $\Re\{s_i\} \leq 0, \forall i$
- Unstable circuit (active circuit):  $\Re\{s_i\} \geq 0, \exists i$

# Zero Natural Frequencies

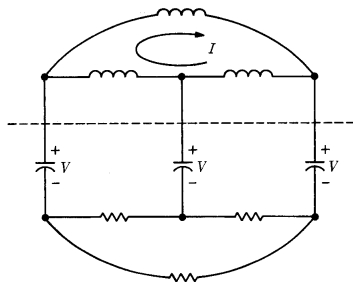


Figure: Zero natural frequency in capacitive cut sets and inductive loops of a passive circuit.

- **Transfer function denominator factor:**  $s$
- **Time-domain behavior:** 1
- **Zero natural frequency creation:** Capacitive cut set, inductive loop, dependent sources (active circuit)
- **Voltage and current of basic elements:** May have different zero natural frequencies

# Natural Frequencies of a Network

# Natural Frequencies of a Network

## Definition (Natural Frequencies of an LTI Network)

$s_i$  is a natural frequency of an LTI network if  $s_i$  is a natural frequency of some voltage or a natural frequency of some current in the network.

## Theorem (Calculation of the Natural Frequencies of a Network)

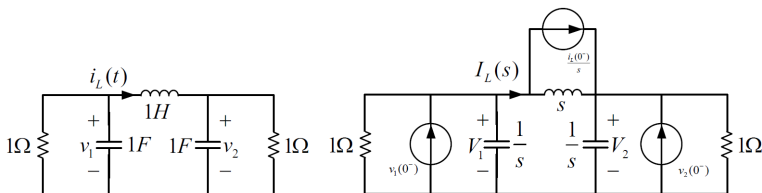
*The nonzero natural frequencies of any linear time-invariant network are identical to the nonzero roots of the equation  $\Delta(s) = \det[\mathbf{P}(s)] = 0$ , where  $\mathbf{P}(s)$  is the matrix of any system of differential equations which describe the network.*

- **Admittance matrix:**  $\mathbf{Y}_n(s)\mathbf{E} = \mathbf{I}_0 + \mathbf{I}_s \Rightarrow \Delta_n(s) = \det[\mathbf{Y}_n(s)]$
- **Impedance matrix:**  $\mathbf{Z}_m(s)\mathbf{I} = \mathbf{E}_0 + \mathbf{E}_s \Rightarrow \Delta_m(s) = \det[\mathbf{Z}_m(s)]$
- **State matrix:**  $(s\mathbf{I} - \mathbf{A})\mathbf{X} = \mathbf{X}_0 + \mathbf{B}\mathbf{W} \Rightarrow \Delta(s) = \det[s\mathbf{I} - \mathbf{A}]$
- **Zero natural frequencies:** Zero eigen values of  $\mathbf{A}$  in state equations
- **Zero natural frequencies:** Capacitive cut sets and inductive loops in a circuit without dependent sources
- **Number of natural frequencies of a network:** Circuit order

# Natural Frequencies of a Network

## Example (Node analysis)

The circuit below has three simple natural frequencies.



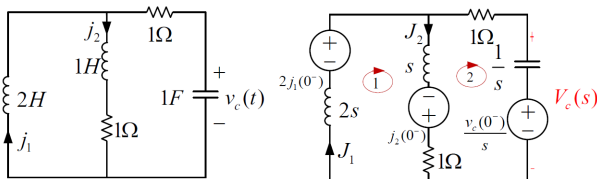
$$\begin{bmatrix} s+1+\frac{1}{s} & -\frac{1}{s} \\ -\frac{1}{s} & s+2+\frac{1}{s} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} v_1(0^-) - \frac{i_L(0^-)}{s} \\ v_2(0^-) + \frac{i_L(0^-)}{s} \end{bmatrix} \Rightarrow \begin{cases} V_1 = \frac{(s^2+s+1)v_1(0^-)+v_2(0^-)-(s+1)i_L(0^-)}{(s+1)(s^2+s+2)} \\ V_2 = \frac{(s^2+s+1)v_2(0^-)+v_1(0^-)+(s+1)i_L(0^-)}{(s+1)(s^2+s+2)} \\ I_L = \frac{V_1-V_2}{s} + \frac{i_L(0^-)}{s} = \frac{v_1(0^-)-v_2(0^-)+(s+1)i_L(0^-)}{s^2+s+2} \end{cases}$$

$$\Delta_n(s) = \det[\mathbf{Y}_n(s)] = \frac{(s+1)(s^2+s+2)}{s} = 0 \Rightarrow s = -1, \frac{-1 \pm j\sqrt{7}}{2}$$

# Natural Frequencies of a Network

## Example (Mesh analysis)

The circuit below has one simple and two repeated natural frequencies.



$$\begin{bmatrix} 3s + 1 & -s - 1 \\ -s - 1 & s + 2 + \frac{1}{s} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 2j_1(0^-) + j_2(0^-) \\ -j_2(0^-) - \frac{v_c(0^-)}{s} \end{bmatrix} \Rightarrow \begin{cases} I_1 = \frac{2(s+1)j_1(0^-) + j_2(0^-) - v_c(0^-)}{(s+1)(2s+1)} \\ I_2 = \frac{2s(s+1)j_1(0^-) 2s^2 j_2(0^-) - (3s+1)v_c(0^-)}{(s+1)^2(2s+1)} \\ V_c = \frac{I_2}{s} + \frac{v_c(0^-)}{s} = \frac{2(s+1)j_1(0^-) - 2sj_2(0^-) - (3s+5)v_c(0^-)}{(s+1)^2(2s+1)} \end{cases}$$

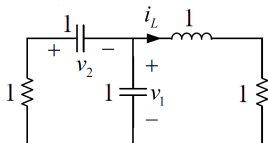
$$\Delta_m(s) = \det[\mathbf{Z}_m(s)] = \frac{(s+1)^2(2s+1)}{s} = 0 \Rightarrow s = -1, -1, -0.5$$



# Natural Frequencies of a Network

## Example (State analysis)

The circuit below has three repeated natural frequencies.



$$\mathbf{X}(t) = \begin{bmatrix} v_1(t) \\ v_2(t) \\ i_L(t) \end{bmatrix}, \mathbf{X}_0 = \mathbf{X}(0) = \begin{bmatrix} v_1(0^-) \\ v_2(0^-) \\ i_L(0^-) \end{bmatrix} \Rightarrow \frac{d\mathbf{X}}{dt} = \begin{bmatrix} -1 & -1 & -1 \\ -1 & -1 & 0 \\ 1 & 0 & -1 \end{bmatrix} \mathbf{X}$$

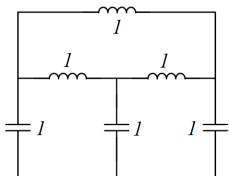
$$s\mathbf{I} - \mathbf{A} = \begin{bmatrix} s+1 & 1 & 1 \\ 1 & s+1 & 0 \\ -1 & 0 & s+1 \end{bmatrix} \Rightarrow \Delta(s) = \det[s\mathbf{I} - \mathbf{A}] = -[0 - (s+1)] + (s+1)[(s+1)^2 - 1]$$

$$\Delta(s) = (s+1)^3 \Rightarrow s = -1, -1, -1$$

# Natural Frequencies of a Network

## Example (Number of network natural frequencies)

The circuit below has order 6, so it has 6 natural frequencies including two zero natural frequencies and 4 repeated imaginary conjugate natural frequencies.



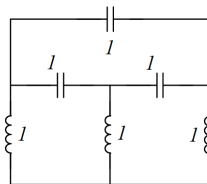
$$\mathbf{Y}_n(s) = \begin{bmatrix} s + \frac{1}{s} + \frac{1}{s} & -\frac{1}{s} & -\frac{1}{s} \\ -\frac{1}{s} & s + \frac{1}{s} + \frac{1}{s} & -\frac{1}{s} \\ -\frac{1}{s} & -\frac{1}{s} & s + \frac{1}{s} + \frac{1}{s} \end{bmatrix}$$

$$\Delta_n(s) = \det[\mathbf{Y}_n(s)] = \frac{(s^2 + 3)^2}{s} = 0 \Rightarrow s = -j\sqrt{3}, +j\sqrt{3}, -j\sqrt{3}, +j\sqrt{3}$$

# Natural Frequencies of a Network

## Example (Number of network natural frequencies)

The circuit below has order 4, so it has 4 natural frequencies including 4 repeated imaginary conjugate natural frequencies.



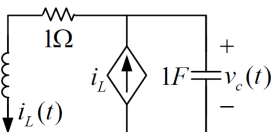
$$\mathbf{Y}_n(s) = \begin{bmatrix} s + s + \frac{1}{s} & -s & -s \\ -s & s + s + \frac{1}{s} & -s \\ -s & -s & s + s + \frac{1}{s} \end{bmatrix}$$

$$\Delta_n(s) = \det[\mathbf{Y}_n(s)] = \frac{(3s^2 + 1)^2}{s} = 0 \Rightarrow s = -j\frac{1}{\sqrt{3}}, +j\frac{1}{\sqrt{3}}, -j\frac{1}{\sqrt{3}}, +j\frac{1}{\sqrt{3}}$$

# Natural Frequencies of a Network

## Example (Number of network natural frequencies)

The circuit below has order 2, so it has 2 natural frequencies including 1 simple natural frequency and 1 simple nonzero real natural frequency.



$$\begin{cases} v_C' = i_L - i_L = 0 \\ i_L' = v_C - i_L \end{cases} \Rightarrow \mathbf{A} = \begin{bmatrix} 0 & 0 \\ 1 & -1 \end{bmatrix}$$

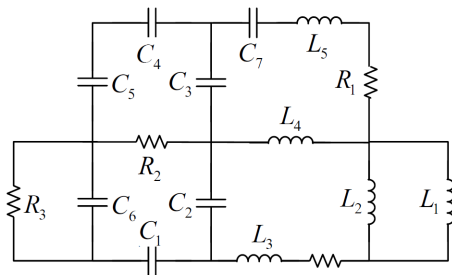
$$\Delta(s) = |s\mathbf{I} - \mathbf{A}| = \begin{vmatrix} s & 0 \\ -1 & s+1 \end{vmatrix} = s(s+1) = 0 \Rightarrow s = 0, -1$$

# Natural Frequencies of a Network

## Example (Number of network natural frequencies)

The circuit below has

- 1 7 capacitors and 5 inductors
- 2 1 independent capacitive loop and 2 independent inductive cut set
- 3 2 independent capacitive cut set and 1 independent inductive loop
- 4  $7 + 5 - 1 - 2 = 9$  natural frequencies and order 9
- 5  $1 + 2 = 3$  zero natural frequencies and  $9 - 3 = 6$  nonzero natural frequencies



# Minimal Differential Equation

## Definition (Minimal Differential Equation)

Minimal differential equation of a variable is the smallest-order differential equation whose corresponding characteristic equation contains all the natural frequencies of the variable.

- **D Operator:**  $Df(t) = \frac{df(t)}{dt}$ ,  $D^{-1}f(t) = \int_{0^-}^t f(\tau)d\tau$
- **Elimination method:** systematic method to obtain minimal differential equation
  - Variable reordering
  - Integral replacement
  - Elementary row operations
  - Upper triangle (diagonal) differential matrix equation
- **Elementary row operations:**
  - Row swap:  $\mathcal{E}_k \leftrightarrow \mathcal{E}_j$
  - Row multiplication:  $\mathcal{E}_k \leftarrow m\mathcal{E}_k$
  - Row sum:  $\mathcal{E}_k \leftarrow \mathcal{E}_k + \mathcal{E}_j$
  - Row mixed operation:  $\mathcal{E}_k \leftarrow m\mathcal{E}_k + p(D)\mathcal{E}_j$

# Minimal Differential Equation

## Example (Minimal differential equation)

The minimal differential equation of the voltage in a zero-input RC circuit is first-order.

$$C \frac{dv(t)}{dt} + \frac{v(t)}{R} = 0, v(0) = V_0 \Rightarrow Cs + R = 0 \Rightarrow s = -\frac{1}{RC} \Rightarrow v(t) = V_0 e^{-\frac{t}{RC}}$$

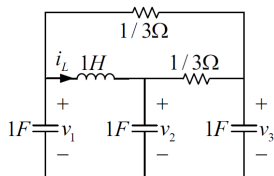
$$C \frac{d^2v(t)}{dt^2} + \frac{1}{R} \frac{dv(t)}{dt} = 0, v(0) = V_0, v'(0) = -\frac{V_0}{RC} \Rightarrow Cs^2 + Rs = 0$$

$$\Rightarrow s = 0, -\frac{1}{RC} \Rightarrow v(t) = K_1 + K_2 e^{-\frac{t}{RC}} \Rightarrow \begin{cases} K_1 + K_2 = V_0 \\ K_2 \frac{-1}{RC} = -\frac{V_0}{RC} \end{cases} \Rightarrow K_1 = 0, K_2 = V_0 \Rightarrow v(t) = V_0 e^{-\frac{t}{RC}}$$

# Minimal Differential Equation

## Example (Minimal differential equation)

The minimal differential equation of  $v_3$  in the circuit below can be obtained using the elimination method.



$$\begin{cases} Dv_1 + D^{-1}(v_1 - v_2) + i_L(0^-) + 3(v_1 - v_3) = 0 \\ Dv_2 + D^{-1}(v_2 - v_1) - i_L(0^-) + 3(v_2 - v_3) = 0 \\ Dv_3 + 3(v_3 - v_1) + 3(v_3 - v_2) = 0 \end{cases}$$

$$\begin{bmatrix} D + 3 + D^{-1} & -D^{-1} & -3 \\ -D^{-1} & D + 3 + D^{-1} & -3 \\ -3 & -3 & D + 6 \end{bmatrix} \begin{bmatrix} v_1(t) \\ v_2(t) \\ v_3(t) \end{bmatrix} = \begin{bmatrix} -i_L(0^-) \\ i_L(0^-) \\ 0 \end{bmatrix}$$

$$\begin{bmatrix} D^2 + 3D + 1 & -1 & -3 \\ -1 & D^2 + 3D + 1 & -3 \\ -3D & -3D & D + 6 \end{bmatrix} \begin{bmatrix} \phi_1(t) \\ \phi_2(t) \\ v_3(t) \end{bmatrix} = \begin{bmatrix} -i_L(0^-) \\ i_L(0^-) \\ 0 \end{bmatrix}$$



# Minimal Differential Equation

## Example (Minimal differential equation (cont.))

The minimal differential equation of  $v_3$  in the circuit below can be obtained using the elimination method.

$$\begin{bmatrix} D + 3 + D^{-1} & -D^{-1} & -3 & -i_L(0^-) \\ -D^{-1} & D + 3 + D^{-1} & -3 & i_L(0^-) \\ -3 & -3 & D + 6 & 0 \end{bmatrix}$$

$$\begin{bmatrix} D^2 + 3D + 1 & -1 & -3 & -i_L(0^-) \\ -1 & D^2 + 3D + 1 & -3 & i_L(0^-) \\ -3D & -3D & D + 6 & 0 \end{bmatrix}$$

$$\begin{bmatrix} -1 & D^2 + 3D + 1 & -3 & i_L(0^-) \\ D^2 + 3D + 1 & -1 & -3 & -i_L(0^-) \\ -3D & -3D & D + 6 & 0 \end{bmatrix}$$

$$\begin{bmatrix} -1 & D^2 + 3D + 1 & -3 & i_L(0^-) \\ 0 & -1 + (D^2 + 3D + 1)(D^2 + 3D + 1) & -3 - 3(D^2 + 3D + 1) & -i_L(0^-) + (D^2 + 3D + 1)i_L(0^-) \\ 0 & -3D - 3D(D^2 + 3D + 1) & D + 6 - 3D(-3) & 0 - 3Di_L(0^-) \end{bmatrix}$$

$$\begin{bmatrix} -1 & D^2 + 3D + 1 & -3 & i_L(0^-) \\ 0 & D^4 + 6D^3 + 11D^2 + 6D & -3D^2 - 9D - 6 & 0 \\ 0 & -3D^3 - 9D^2 - 6D & 10D + 6 & 0 \end{bmatrix}$$

# Minimal Differential Equation

## Example (Minimal differential equation (cont.))

The minimal differential equation of  $v_3$  in the circuit below can be obtained using the elimination method.

$$\begin{bmatrix} -1 & D^2 + 3D + 1 & -3 & i_L(0^-) \\ 0 & -3D^3 - 9D^2 - 6D & 10D + 6 & 0 \\ 0 & D^4 + 6D^3 + 11D^2 + 6D & -3D^2 - 9D - 6 & 0 \end{bmatrix}$$

$$\begin{bmatrix} -1 & D^2 + 3D + 1 & -3 & i_L(0^-) \\ 0 & -3D^3 - 9D^2 - 6D & 10D + 6 & 0 \\ 0 & 0 & -3D^2 - 9D - 6 + (\frac{1}{3}D + 1)(10D + 6) & 0 + (\frac{1}{3}D + 1)0 \end{bmatrix}$$

$$\begin{bmatrix} -1 & D^2 + 3D + 1 & -3 & i_L(0^-) \\ 0 & -3D^3 - 9D^2 - 6D & 10D + 6 & 0 \\ 0 & 0 & \frac{1}{3}D^2 + 3D & 0 \end{bmatrix}$$

$$(\frac{1}{3}D^2 + 3D)v_3(t) = 0 \Rightarrow v_3'' + 9v_3' = 0 \Rightarrow s^2 + 9s = 0 \Rightarrow s = 0, -9$$

# The End